

# Packet-Based Simulation for Optical Wireless Communication

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**Abstract**—This paper presents packet-based simulation tools for free-space-optical (FSO) wireless communication. We implement the well-known propagation models for free-space-optical communication as a set of modules in NS-2. Our focus is on accurately simulating line-of-sight (LOS) requirement for two communicating antennas, the drop in the received power with respect to separation between antennas, and error behavior. In our simulation modules, we consider numerous factors affecting the performance of optical wireless communication such as visibility in the medium, divergence angles of transmitters, field of view of photo-detectors, and surface areas of transceiver devices.

**Index Terms**—Free-space-optics, wireless simulation, FSO propagation

## I. INTRODUCTION AND MOTIVATION

Wireless communication has traditionally been realized via omnidirectional radio frequency. Radio frequency has the major advantage of propagating in all directions enabling a receiver to roam inside the transmission sphere without experiencing a link disruption, although, it may encounter fading and hidden nodes as obstacles hurting the uniformity of the signal and new communicating nodes present in the propagation medium. Nevertheless, a typical RF-enabled node will have a large throughput gap with optical backbone of the network which reveals the last mile problem [1]–[3]. Pushing more aggressive medium access control (MAC) protocols that operate in much finer grained time scales and employing innovative multihop hierarchical cooperative MIMO [4] techniques remedy the issue partially in the cost of increased complexity. Marginal benefit of such approaches have become smaller due to the increased saturation of the RF spectrum. The throughput gap between optical backbone and the wireless last-mile calls for more radical approaches involving wireless spectrum bands physically much larger than the RF.

Free-space-optical (FSO) (i.e., optical wireless) communication provides an attractive approach complementary to the legacy RF-based wireless communication. Most significant difference between FSO and RF is the requirement of line-of-sight in FSO, adding *space-division multiplexing* (i.e., spatial reuse) to already known multiplexing techniques such as wave-length and time division multiplexing. RF suffers from increased power consumption per interface compared to FSO because of the significantly larger volume of medium that needs to be covered by an individual interface. RF-based communication

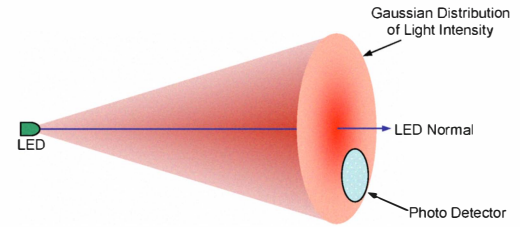


Fig. 1. Gaussian distribution of light intensity at the receiver plane.

also has a greater need to employ complex security protocols to address security concerns that rise because of the higher risk of interception especially in military applications.

A typical FSO transmitter (e.g., LASER, VCSEL or LED) forms a cone shaped volume in 3 dimensions (Figure 1) in which a potential receiver equipped with a photo detector can receive the signal. The exact shape of this cone is determined by the transmission power (for range) and divergence angle. A LASER has the smallest (in micro radian range) and an LED has the widest (a few hundred milli radians) divergence angle of the three types of transmitters. FSO can operate in large swathes of unlicensed spectrum reaching speeds up to ~1 Gbps. Additionally, FSO transceivers have much smaller form factors, are less power-consuming (100 microwatts for 10-100 Mbps), very reliable (lifetime of more than 10 years), cheap and offer highly directional beams for spatial reuse and security.

Simulation efforts of free-space-optical communication have primarily focused on physical propagation models [5], [6]. Researchers also worked on numerical analysis of the wireless optical communication and especially considered error analysis of the channel in extreme scenarios such as atmospheric turbulence [7]–[9]. Our focus is mainly on *packet-based simulation* of free-space-optical wireless communication.

Network Simulator 2 [10] is a widely-used open source discrete event simulation platform for networking research. NS-2 has been developed and maintained by the research community since 1989, letting contributors enhance its capabilities by implementing necessary parts. Hence, the platform allowed researchers to observe many important phenomenons in wireless networking. Our effort is on accurately simulating

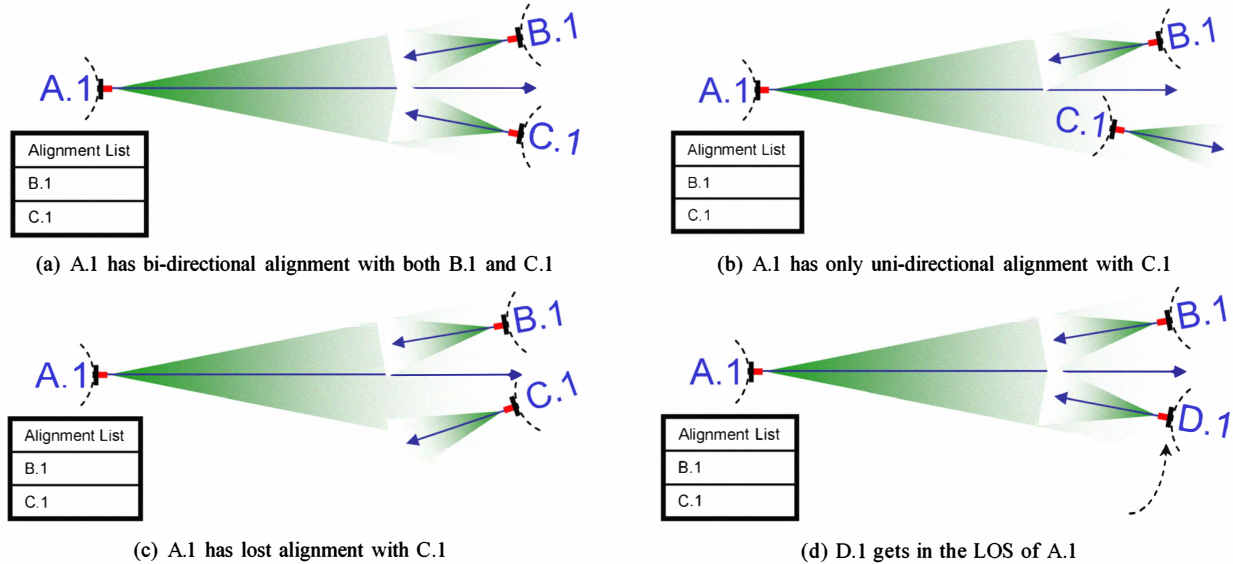


Fig. 2. Types of possible alignment loss/gain during a timer period.

propagation model [11] of FSO communication, line-of-sight (LOS) requirement for two communicating antennas, the drop in the received power with respect to separation between antennas, and error behavior. Our study considers visibility in the medium, divergence angles of transmitters, field of view of photo-detectors, and surface areas of transceiver devices to identify their effect on the communication performance.

Physical free-space-optical propagation model along with directional communication did not exist in NS-2 prior to our contribution. We present a transceiver structure (consisting of an LED and a photo-diode) that has a divergence angle which determines the field of view of the transceiver. The divergence angle of a transceiver is very fundamental to our contribution since it is the main factor that determines if two transceivers are aligned with each other. Moreover, we model the received power as we increase the separation between a transmitter and a receiver, which also affects bit error rate. Additionally, we investigate the effect of visibility on the system. Visibility is a particularly important ingredient since it has conventionally been the most important parameter for designing point-to-point FSO links. Prior to our work, a wireless (RF) link in a packet-based simulator has traditionally been implemented in an omnidirectional way; hence, there was not a way to establish directional links that can use the same frequency band simultaneously without interfering one another. We implemented NS-2 enhancement modules that can:

- Determine the existence of directional links between transceivers of different nodes and deliver packets accordingly,
- Mimic the characteristics of an FSO link in power reception, noise, bit error rate profiles.

In Section II, we present the well-known theoretical model for FSO propagation in a non-turbulent medium. We give the details of our NS-2 implementation in Section III. Section IV provides the results of our experiments to show the power, BER and error probability behavior from our FSO simulation

modules. Lastly, we summarize our work in Section V.

## II. THEORETICAL FSO PROPAGATION MODEL

We used well-known FSO propagation models [11] to simulate power attenuation characteristics of an FSO signal. LEDs' light intensity profile follows the Lambertian law [11], i.e., intensity is directly proportional to the cosine of the angle from which it is viewed. At a distance  $Z$ , let the received power along the beam be  $P_Z$ . Based on the Lambertian law, at an arbitrary angle  $\alpha$  from the vertical axis and at a distance  $Z$ , the intensity would be:  $P_{\alpha,Z} = P_Z \cos(\alpha)$ . For edge-emitting LEDs, this is improved by a factor  $u$  in the power of cosine, i.e. the intensity is given by:  $P_{\alpha,Z} = P_Z \cos^u(\alpha)$ .

Also, as a generic definition for all FSO transmitters, the beam radius  $w_Z$  at the vertical distance  $Z$  is defined as the radial distance at which the received power is  $\frac{1}{e^2} P_Z$ . So, the divergence angle  $\theta$  is the special value of  $\alpha$ , where the ratio  $P_{\alpha,Z}/P_Z = 1/e^2$  holds, which means  $\theta$  can be calculated by  $\theta = \tan^{-1}(w_Z/Z)$ .

FSO propagation is affected by both the atmospheric attenuation  $A_L$  and the geometric spread  $A_G$ , which practically necessitates the source power to be greater than the power lost. The *geometric attenuation*  $A_G$  is a function of transmitter radius  $\gamma$ , the radius of the receiver (on the other receiving FSO node)  $\varsigma$  cm, divergence angle of the transmitter  $\theta$  and the distance between the transmitting node and receiving node  $R$ :

$$A_G = 10 \log \left( \frac{\varsigma}{\gamma + 200R\theta} \right)^2$$

The *atmospheric attenuation*  $A_L$  consists of absorption and scattering of the laser light photons by the different aerosols and gaseous molecules in the atmosphere. The power loss due to atmospheric propagation is given by Bragg's Law [11] as:

$$A_L = 10 \log(e^{-\sigma R})$$

where  $\sigma$  is the attenuation coefficient consisting of atmospheric absorption and scattering. For the wavelengths used for FSO

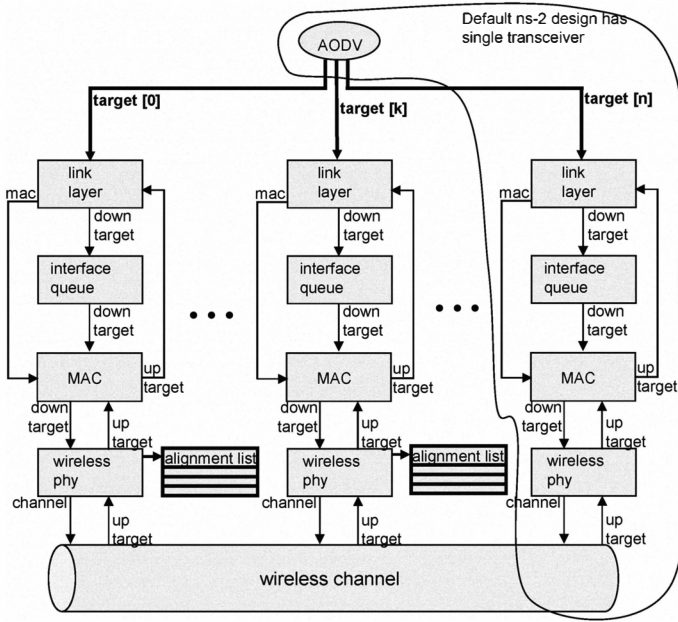


Fig. 3. FSO node structure with a separate stack for each optical transceiver. AODV is modified so that it is capable of handling multiple network interfaces. WirelessPhy is also modified to keep a list of aligned transceivers.

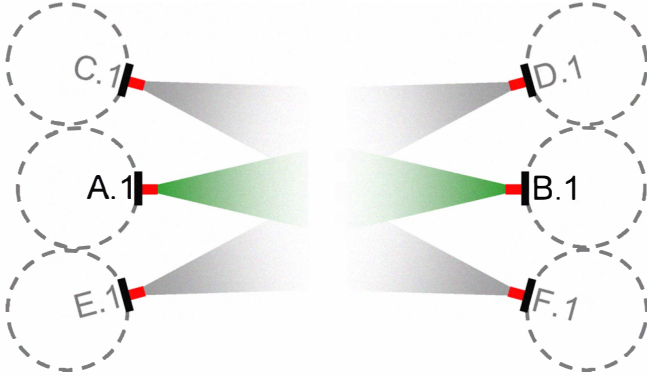


Fig. 4. Noise in FSO transmission: transceivers C.1, D.1, E.1 and F.1 contribute to the noise for the communication between A.1 and B.1.

communication, Mie scattering dominates the other losses, and therefore is given by [12]:

$$\sigma = \frac{3.91}{V} \left( \frac{\lambda}{550} \right)^{-q}.$$

In the above formulation of  $\sigma$ ,  $V$  is the atmospheric visibility in kilometers,  $q$  is the size distribution of the scattering particles whose value is dependent on the visibility:

$$q = \begin{cases} 1.6 & V \geq 50\text{km} \\ 1.3 & 6\text{km} \leq V < 50\text{km} \\ 0.583V^{1/3} & V < 6\text{km} \end{cases}$$

### III. IMPLEMENTATION IN NETWORK SIMULATOR 2

Our contribution (Figure 3) includes a full implementation of FSO propagation model to calculate source and reception power

TABLE I  
TABLE OF DEFAULT VALUES COMMON TO EACH SIMULATION SET IN OUR EXPERIMENTS.

Parameter Name	Default Value
Visibility	6 km
Number of interfaces	8
Transmission range and separation between nodes	30 m
Divergence angle	1 rad
Photo detector diameter	5 cm
LED diameter	0.5 cm
Per-bit error probability	$10^{-6}$
Noise	$1.1428\text{e-}12$ Watt
Capture threshold	$1.559\text{e-}11$ Watt
Receive threshold	$3.652\text{e-}10$ Watt

of packets under relevant parameters such as atmospheric attenuation, visibility, Gaussian-distributed geometric beam spread (Figure 1), photo-detector threshold, transmitter and receiver diameters, divergence angle, desired error probability per bit and noise (Table I). We use all the above parameters to determine the reception power of a transmission using the theoretical models discussed in Section II. We also take noise into consideration (Figure 4) while determining signal to noise ratio. The noise in FSO is inherently different from RF in the sense that it is directional.

The directional FSO antenna model that we used has 3-D pointing and divergence angle features, as well as diameters of LED/transmitter and photo-detector components. The light beam forms a cone shape in 3-D (Figure 1) as it propagates away from the source. Divergence angle of the transmitting LED dictates the shape of propagation. We use a Gaussian distribution of light intensity when considering a cross cut of this cone. On the receiving side, the photo detector also has a field of view which is assumed to be the same with LED's in the transceivers we simulate.

At a given time, the transceivers in the system form such directional optical links. Those links stay unchanged as long as there is no mobility of either end. With mobility involved, each transceiver can be aligned or can get misaligned to a number of other transceivers. To keep track of such alignment and misalignment events, we implemented a timer mechanism for periodic checking and establishment of LOS alignment lists for each transceiver. We use a new alignment-table-based channel model for delivering packets only to the candidate receiver antennas that reside in the transmitter's alignment list.

Whenever the channel chooses to deliver a packet to a receiver, we take the transmission power and spread it in a Gaussian manner onto a circular area which makes the cross-cut of the illumination cone (Figure 1). Then, we calculate the amount of light that drops on the surface of the receiver using its diameter, its separation from the transmission normal and the angle it makes with the transmission normal. If the received power is greater than carrier sense threshold, then the packet is considered for noise for the currently received packet, if there is any. If the power is greater than receive threshold, then it is considered for reception. After deciding the received power level, we need to determine if the packet is erroneous. We take the reception power of the packet and calculate the theoretical



Received Power Distribution in 2-D

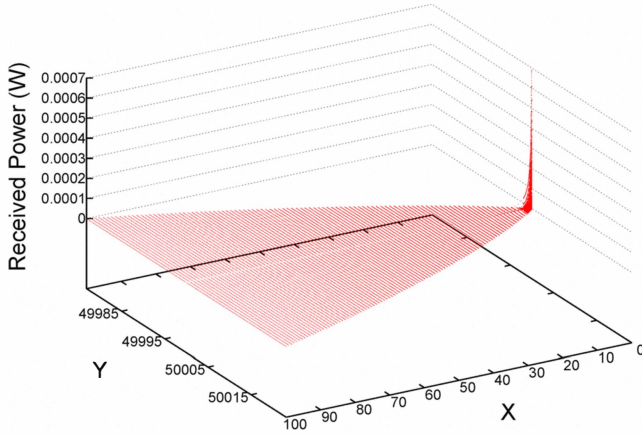


Fig. 5. Received power in the field of view of a 1 rad light source.

Received Power Distribution in 2-D

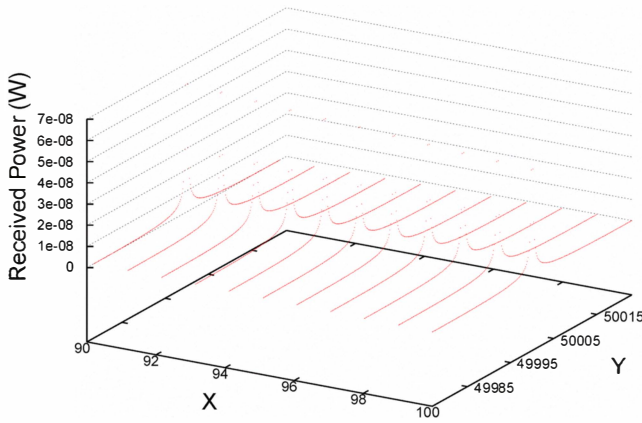


Fig. 6. Received power between 90 and 100 meter ranges.

bit error probability using the visibility in the medium, distance between transmitter and receiver and noise. From this bit error probability we calculate the probability that the whole packet can be received without any bit errors. Lastly, drawing a uniform decides if this packet should be captured without any errors or contribute to the noise.

#### A. Alignment Lists and Alignment Timer

We implemented a timer mechanism in NS-2 that goes off every half-a-second (which can be tuned) and determines the alignments among the transceivers. This timer mechanism corresponds to “automatically” re-checking availability of LOS alignment. An ongoing transmission may experience a disconnection due to mobility, sway or vibration of either nodes [13]–[15]. In such a disconnection, automatic alignment checking can be considered as the “search” phase before starting to send data. The search phase discovers possible alignment establishments which are discovered via the alignment timers in our simulations. In simulation scenarios with high mobility rates, the alignment timer could be much longer and coarse

Distance vs Theoretical Bit Error Probability

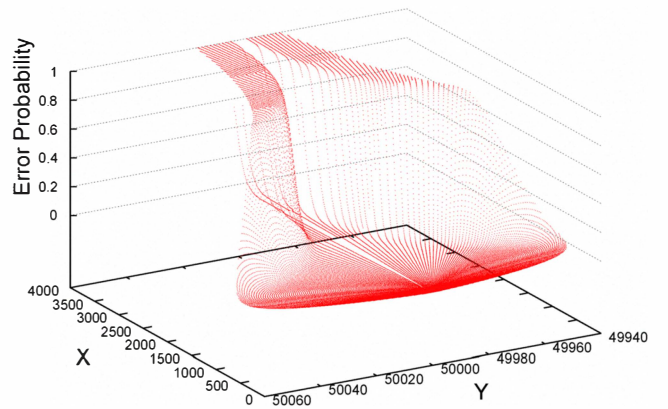


Fig. 7. Probability of error increases as a receiver is moved away from the transmitter.

for alignment detection and establishment. Hence, the mutual alignment between two transceivers might not be preserved during a complete alignment period, a situation which needs to be carefully modeled in the simulation setup. Once the alignment timer expires, it takes one primary transceiver at a time and creates a list of candidate transceivers that both the primary transceiver and candidate transceiver are in each others’ line-of-sight, hence the term mutual alignment.

Figure 2 depicts the set of possible events that may occur before the alignment timer goes off in a scenario with multiple transceivers each from different nodes (A, B, C and D) with only their first transceiver shown and from A.1’s perspective. In the simplest case, alignments can stay unmodified like in Figure 2(a). In Figure 2(b), we see that node C moved and its transceiver C.1 can not see transceiver A.1 any more. But, transceiver A.1 can still see C.1 and because the alignment timer has not fired yet, A.1 continues to keep an entry for C.1 in its alignment list thinking that it is still aligned. Notice that, if the alignment timer expires in such a case, C.1 will not be placed in A.1’s list since the alignment between the two is lost and not *mutual*. That is, in our simulations the alignment is “bi-directional” and both A.1 and C.1 should see each other in order for communication to take place. Note that this is a conservative assumption for line-of-sight establishment and there is still room for improvement.

For the third case in Figure 2, C.1 might have turned its back or just moved out of line-of-sight of A.1. Hence both have lost alignment with each other and although they will continue to keep entries for each other packets will be dropped until the alignment timer expires and the alignments are re-established through other transceivers or paths.

The fourth case in Figure 2 is a new transceiver, D.1, gets in the LOS of A.1. However, D.1 and A.1 will not be able to exchange data packets until the alignment timer goes off again and the alignment lists are updated. This is another major conservative assumption in our simulations and is true; regardless of the alignment’s nature, uni-directional or bi-

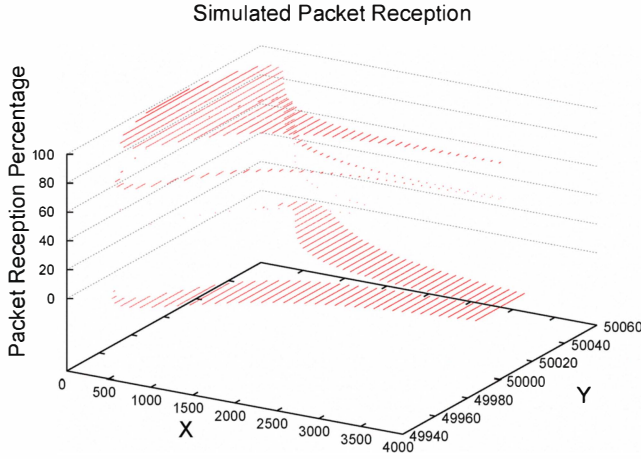


Fig. 8. Percentage of successfully delivered packets decreases as the receiver is moved away from the light source. Used transport agent is UDP.

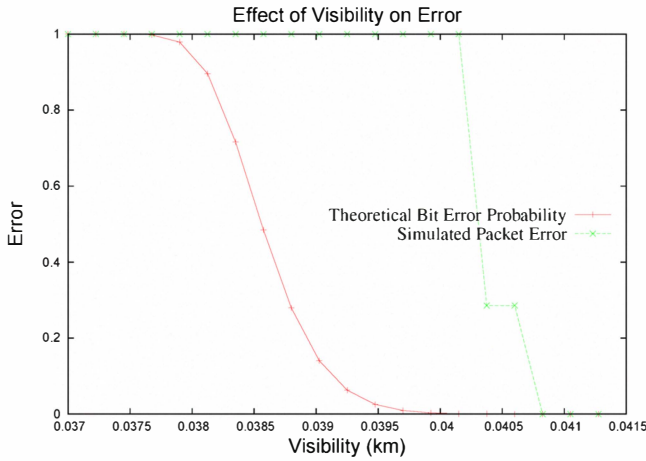


Fig. 9. Probability of error decreases as the visibility in the medium is increased. Percentage of delivered packets follows a similar but coarser grained behavior.

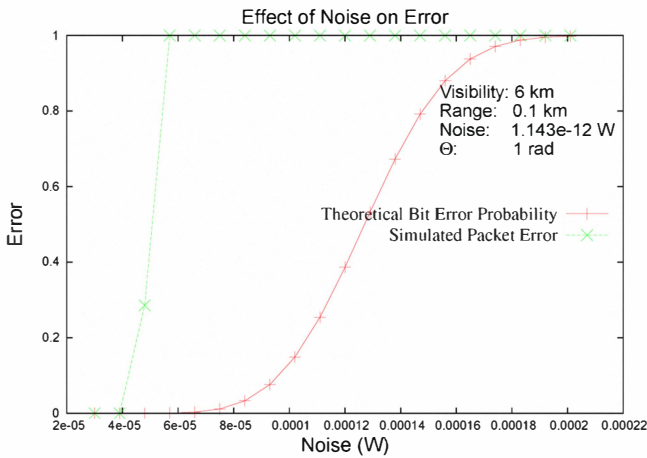


Fig. 10. Theoretical error probability and simulated packet error increase as the noise is increased.

directional. If D.1 keeps staying in LOS of A.1, new entries<sup>5</sup> will be created for each other in their alignment lists when the alignment timer expires. Only after then, the two transceivers will be able to exchange packets.

#### IV. RESULTS AND COMPARISON

To show that our FSO simulation modules comply with the theoretical propagation model, we have done several simulation experiments. Our experiments involved two transceivers positioned in different ways with respect to each other. We observed received power, error probability and bit error rate in packet transmissions while varying important parameters like the separation between the two transceivers, visibility and noise.

##### A. Effect of Separation in Received Power, Theoretical Bit Error Probability and Simulated Packet Error

Complying with theoretical framework, our results reveal that the received power follows Lambertian law [11] from the transmitter itself and normal of the transmitter as depicted in Figure 5. Original transmission power for this scenario is calculated for 0.1 meter. We increased separation between transmitter and receiver antennas from 0.01 meter to 100 meters in our simulations (Figure 5). Figure 6 shows the Gaussian distribution of the received light intensity clearly as the receiver is moved away from transmitter's normal line by focusing on the last 10 meters of Figure 5.

Distance also affects theoretical error probability and simulated packet error since the received power decreases significantly. We sampled theoretical error with separation between antennas ranging from 10 meters to 4000 meters. Figure 7 shows that the theoretical error probability increases significantly as the receiver is moved away from the transmitter while keeping the transmission power same. Similarly, simulated packet error is shown in Figure 8 which follows theoretical error probability.

##### B. Effect of Visibility in Theoretical Bit Error Probability and Simulated Packet Error

Low visibility in the medium makes the light experience more deviation from its intended direction by hitting aerosols in the air. This causes the received light intensity to drop which causes more bit errors. Hence, increasing visibility decreases theoretical error probability and simulated packet error. For this simulation scenario, the power is calculated for 100 meters with 6 km visibility and kept the same for all the simulations. Separation between antennas is 100 meters. We increased visibility from 0.037 km to 0.041275 km. In Figure 9, we show that the visibility in the medium affects theoretical bit error probability and simulated packet error significantly. From the figure, we can see that if visibility is set to a value from 0 to 0.037 km, the system experiences a high level of error and after 0.04 km, it recovers.

### C. Effect of Noise in Theoretical Bit Error Probability and Simulated Packet Error

We found that noise has an important impact on theoretical bit error probability and simulated packet error since it will become harder for the receiver to operate at a low signal-to-noise ratio. We used a transmission power that reaches 100 meters with a noise level of  $1.1428 \times 10^{-12}$  Watt for all of our simulations in this scenario. We increased the noise in the medium from  $3.0 \times 10^{-5}$  W to  $2.01 \times 10^{-4}$  W and found that both the theoretical error probability and simulated packet error are increased considerably as depicted in Figure 10.

### V. SUMMARY

In this paper, we presented our contribution to NS-2 in simulating free-space-optical links. We took visibility in the medium, divergence angles of transmitters, field of view of photo-detectors, and surface areas of transceiver devices into account while implementing such enhancements. We provided results of our efforts to comply with theoretical models, showing drop in received power, theoretical error probability and simulated packet error with respect to separation, medium visibility and noise.

### ACKNOWLEDGMENT

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### REFERENCES

- [1] NRC Committee on Broadband Last Mile Technology, *Broadband: Bringing Home the Bits*. National Research Council, CSTB, 2002.
- [2] A. R. Moral, P. Bonenfant, and M. Krishnaswamy, "The optical internet: architectures and protocols for the global infrastructure of tomorrow," *IEEE Communications Magazine*, vol. 39, no. 7, pp. 152–159, 2001.
- [3] D. K. Hunter and I. Andonovic, "Approaches to optical internet packet switching," *IEEE Communications Magazine*, vol. 38, no. 9, pp. 116–122, 2000.
- [4] A. Ozgur, O. Levesque, and D. Tse, "Hierarchical cooperation achieves optimal capacity scaling in ad hoc networks," *IEEE Transactions on Information Theory*, vol. 53, no. 10, pp. 3549–3572, February 2007.
- [5] C. R. Lomba, R. T. Valadas, and A. M. de Oliveira Duarte, "Efficient simulation of the impulse response of the indoor wireless optical channel," *Proc. of the Seventh IEEE International Symposium on Personal, Indoor and Mobile Communications*, pp. 257–261, 1996.
- [6] J. B. Joseph, J. R. Barry, J. M. Kahn, W. J. Krause, E. A. Lee, and D. G. Messerschmitt, "Simulation of multipath impulse response for indoor wireless optical channels," *IEEE Journal on Selected Areas in Comm*, vol. 11, pp. 367–379, 1993.
- [7] M. Safari and M. Uysal, "Relay-assisted free-space optical communication," Nov. 2007, pp. 1891–1895.
- [8] S. Navidpour, M. Uysal, and M. Kavehrad, "BER performance of free-space optical transmission with spatial diversity," *Wireless Communications, IEEE Transactions on*, vol. 6, no. 8, pp. 2813–2819, August 2007.
- [9] R. Valadas, "Interference modelling for the simulation of ieee 802.11 infrared local area networks," vol. 1, Oct 1996, pp. 257–261 vol.1.
- [10] "The Network Simulator," <http://www.isi.edu/nsnam/ns/>.
- [11] H. Willebrand, *Free Space Optics*. Sams Pubs, 2001, 1st Edition.
- [12] H. C. Van de Hulst, *Light Scattering by Small Particles*. John Wiley and Sons, 1957.
- [13] M. Yuksel, J. Akella, S. Kalyanaraman, and P. Dutta, "Free-space-optical mobile ad hoc networks: Auto-configurable building blocks," *ACM Springer Wireless Networks*, vol. 15, no. 3, pp. 295–312, April 2009.
- [14] M. Bilgi and M. Yuksel, "Multi-element free-space-optical spherical structures with intermittent connectivity patterns," in *Proceedings of IEEE INFOCOM Student Workshop*, 2008.
- [15] B. Nakhkoob, M. Bilgi, M. Yuksel, and M. Hella, "Multi-transceiver optical wireless spherical structures for manets," *IEEE Journal on Selected Areas of Communications*, vol. 27, no. 9, 2009.